# Polytopic Linear Models-Based Output Tracking Control of a Single-Link Flexible Joint Robot Manipulator 

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#### Abstract

In this paper, to solve the output tracking problem of a single-link flexible joint manipulator, Polytopic Linear Models (PLMs) of the dynamics are made to take advantage of this method. Although linear control methods are very useful due to their powerful theories and simplicity, they can only be used in a neighborhood of the equilibrium point. One way to solve this problem is a PLMs-based method that linearizes the dynamics around several operating points. Therefore, in this paper, after calculating the PLMs of the manipulator, a state feedback control is applied to the derived linear dynamics that are augmented with the dynamics of the output tracking error. An extended method is used to decompose the scheduling space to construct PLMs, which is the segregation method improved with an extra aggregation. In order to avoid creating a large number of local models, an axis-oblique decomposition strategy is used instead of an axis-orthogonal decomposition. In addition, the scheduling functions of the PLMs are determined such that overlaps between the regions are avoided. By this selection, the output tracking problem becomes as a Linear Matrix Inequality (LMI) problem instead of a bilinear matrix inequality problem, which is more difficult to solve and may not lead to an optimal global solution.


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## 1- Introduction

In modern manipulators, flexibility is unavoidable due to the requirements of industrial automation. Flexible links and flexible joints are found in many applications, such as servicing sector, building and maintenance various space stations, gantry cranes, atomic force microscopes, and medical and defense industries [1-3].

Generally, flexibility is an undesirable feature in manipulators since significant control problems like vibrations and deflections are accrued [1, 4-6]. Therefore, the development of control methods for flexible structures has received special attention from researchers [7, 8].

Different control approaches have been discussed to control flexible joint manipulators in literature. The methods such as LQR [9, 10], PID [3, 11], and state feedback [3, 12] are linear model dynamics based control methods. Additionally, adaptive control [13, 14], fuzzy model-based control [15, 16], sliding mode control methods [17-19], and neuro controllers [20,21] are nonlinear model dynamics-based methods which have been used to control the flexible joint manipulator. Linear systems-based analysis and design methods including controllability, observability, stability analysis, and controller design are well understood and well developed compared to nonlinear systems analysis and design methods. However, the drawback is that the results are just locally valid. Therefore, describing nonlinear dynamical systems by combinations of
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linear sub-systems has received significant attention to have a more accurate model. For example, Linear Parameter Varying (LPV) model is used for control purposes, including manipulator control problems [22-24]. However, in this paper we use a simpler method (Polytopic Linear Models (PLMs)) for a flexible joint robot, since in this model the existence of the time-varying parameters complicates the system analysis and controller design. PLMs are built upon a number of linear models introduced by [25]. These linear models describe the system locally within a so-called scheduling regime. Furthermore, even with the existence of uncertainties, the stability of the nonlinear system is guaranteed if its approximated dynamics PLMs is stable [25].

In this paper, PLMs are used to model and control a sin-gle-link flexible joint manipulator. In PLMs approximation, scheduling functions determine forms of scheduling regimes. We select them to avoid interference between regions, leading to the tracking control problem becoming an LMI problem rather than a bilinear matrix inequality problem. In addition, to avoid creating many local models, an axis oblique decomposition strategy is used instead of an axis-orthogonal decomposition strategy.

The rest of the paper is organized as follows. Section 2 presents the dynamical model of a single-link flexible manipulator. In section 3, approximation by the PLMs method, including determination of the scheduling regimes and calculating the PLMs parameters, are discussed. Section 4 presents


Fig. 1. A single-link flexible joint manipulator [3].
the stability analysis and the method of output tracking control for the flexible joint manipulator. The simulation results demonstrate the effectiveness of the method. Finally, the conclusion of the paper is provided in section 5 .

## 2- Dynamical Model of the Single-Link Flexible Joint Manipulator

Consider the single joint flexible manipulator shown in Fig. 1. The joint mounted on the shaft moves according to the rotation direction of the motor, where $\theta$ is the rotation angle of the joint and $\alpha$, is the oscillation angle of the end effectors.

A mathematical model for this manipulator is obtained from Lagrange equations, as follows [3]:

$$
\begin{align*}
& \dot{x}_{1}=x_{3} \\
& \dot{x}_{2}=x_{4} \\
& \dot{x}_{3}=\frac{K_{s}}{J_{h}} x_{2}-\frac{K_{m}^{2} K_{g}^{2}}{J_{h} R_{m}} x_{3}+\frac{K_{m} K_{g}}{J_{h} R_{m}} u  \tag{1}\\
& \dot{x}_{4}=-\left(\frac{K_{s}}{J_{h}}+\frac{K_{s}}{J_{l}}\right) x_{2}+\frac{K_{m}^{2} K_{g}^{2}}{J_{h} R_{m}} x_{3}+\frac{m g h}{J_{l}} \sin \left(x_{1}+x_{2}\right)-\frac{K_{m} K_{g}}{J_{h} R_{m}} u \\
& y=x_{1}+x_{2} .
\end{align*}
$$

The dynamics consist of four state variables. The first two are $x_{1}=\theta$, and $x_{2}=\alpha$, the others are their derivations $x_{3}=\dot{\theta}$, and $x_{4}=\dot{\alpha} . u$ is the system input which is the input voltage. The output is the angle of the end effector, which is the sum of the two angles of the rotation angle of the joint $\theta$ and the oscillation angle of the link $\alpha$. The values of the parameters are given in Table 1 [3].

To use the dynamics (1) in a simple form, it is rewritten as follows:

$$
\begin{align*}
\dot{x} & =f(x)+b u \\
y & =c x \tag{2}
\end{align*}
$$

where
$f(x)=\left[x_{3}, \quad x_{4}, \frac{K_{s}}{J_{h}} x_{2}-\frac{K_{m}^{2} K_{g}^{2}}{J_{h} R_{m}} x_{3}, \quad-\left(\frac{K_{s}}{J_{h}}+\frac{K_{s}}{J_{t}}\right) x_{2}+\frac{K_{m}^{2} K_{g}^{2}}{J_{h} R_{m}} x_{3}+\frac{m g h}{J_{t}} \sin \left(x_{1}+x_{2}\right)\right]^{T}$,
and
$b=\left[\begin{array}{llll}0, & 0, & \frac{K_{m} K_{g}}{J_{h} R_{m}}, & -\frac{K_{m} K_{g}}{J_{h} R_{m}}\end{array}\right]^{T}$,
$c=\left[\begin{array}{llll}1, & 1, & 0, & 0\end{array}\right]$.

## 3- Approximation by PLMs

In this section, an $\varepsilon$-accurate PLMs for the flexible joint manipulator dynamics (1) is derived. The PLMs have the following form [25]:

$$
\begin{align*}
& \dot{x}=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left(A_{i} x+a_{i}\right)+b u,  \tag{3}\\
& y=c x
\end{align*}
$$

where $A_{i} \mathrm{~s}$ and $a_{i} \mathrm{~s}$ are the parameters of the PLMs with the size of $4 \times 4$, and $4 \times 1$, respectively. $w_{i}\left(Z^{i}\right) \mathrm{s}$ are the scaler scheduling functions, $Z^{i}$ is the $i^{t h}$ scheduling regime, and $N$ is the number of linear models or scheduling regimes. The approximated model has to be close to the nonlinear dynamical model in the sense that they show the same behaviors. One criterion for measuring the accuracy is the Euclidean distance between these two models. The criterion is defined as follows:

$$
\begin{equation*}
\operatorname{diff}_{f g}(x)=\sup \left\|f(x)-\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left(A_{i} x+a_{i}\right)\right\|_{2} \leq \varepsilon, \tag{4}
\end{equation*}
$$

where $\varepsilon$ denotes the approximation accuracy.

## 3-1- Scheduling Regimes

According to the degree of nonlinearity of the system model, the number of local linear models and their regions forming result in the desired accuracy, should be determined.

In order to reduce the number of linear models or scheduling regimes (denoted by $N$ ), the dimension of the scheduling space (denoted by $Z$ ) compared to the operating space should be reduced as much as possible since $N$ grows exponentially with $Z$. The main factor for reducing the dimension of the scheduling space is the existence of some state variables that the dynamical equations are linear with respect to them. In this case, these variables do not participate in the partition-

Table 1. Parameters of the flexible joint manipulator.

| Symbol | Description | Value |
| :---: | :---: | :---: |
| $J_{l}$ | Inertia of flexible manipulator | $0.003882 \mathrm{Kgm}^{2}$ |
| $R_{m}$ | Motor resistance | $15.5 \Omega$ |
| $K_{g}$ | Gear ratio of redactor | 1/36 |
| $K_{m}$ | Motor constant | 0.0089 Ns.rad ${ }^{-1}$ |
| $K_{s}$ | Flexibility coefficient of joint | $5.468 \mathrm{Nm}^{-1}$ |
| $m$ | Mass of the flexible joint | 0.03235 Kg |
| $g$ | Gravitational acceleration | $-9.81 \mathrm{NKg}^{-1}$ |
| $h$ | Distance to center of gravity of rotational platform of flexible manipulator | 0.06 m |
| $J_{h}$ | Inertia of rotational platform | $0.00035 \mathrm{Kgm}^{2}$ |

ing, since the approximation of the relevant terms becomes themselves by any partitioning. For the flexible joint manipulator dynamics (1), all terms of the first, second, and third equations are linear with respect to the state variables and the input. For the fourth equation, the term $\sin \left(x_{1}+x_{2}\right)$ is nonlinear, and therefore it is enough to partition only this term for partitioning the scheduling space of the system (1). Afterwards, parameters of the following approximation should be calculated for every region:

$$
\begin{equation*}
\sin \left(x_{1}+x_{2}\right) \cong \sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left(\alpha_{i} x_{1}+\beta_{i} x_{2}+\lambda_{i}\right), \tag{5}
\end{equation*}
$$

where $\alpha_{i}$ and $\beta_{i}$ are the entries of the fourth row, the columns one and two in the matrix $A_{i}$, respectively. $\lambda_{i}$ is the fourth element of the vector $a_{i}$.

To decompose the scheduling space $Z$ in $N$ disjoint regions $Z^{i} \mathrm{~s}$, various procedures like uniform decomposition, aggregation, segregation, and extended method that is segregation with an extra aggregation are used [25]. The methods
should be applied such that $\varepsilon$-accuracy is achieved. In addition, there are two major strategies for partitioning the scheduling space: an axis-orthogonal partitioning strategy [25] and an axis-oblique partitioning strategy [26]. The advantages of the first strategy include efficient structure and using parameters learning schemes such as LOLIMOT (local linear model tree). However, the axis-orthogonal partitioning restricts the model flexibility. The second strategy, an axis oblique decomposition strategy, grasps the character of the nonlinearity and therefore requires significantly less number of linear models. This paper reduces the number of models $N$, using the extended method with an axis-oblique decomposition strategy instead of the axis-orthogonal decomposition strategy.

To approximate $\sin \left(x_{1}+x_{2}\right)$, we define a variable as follows for simplicity and to avoid creating a large number of hyper cubes and then partition $\sin (\psi)$ with respect to this new variable:

$$
\begin{equation*}
\psi \triangleq x_{1}+x_{2}, \tag{6}
\end{equation*}
$$



Fig. 2. Partitioning $\sin (\psi)$ by segregation method.


Fig. 3. Segregation regions of $\sin (\psi)$ versus the corresponding values $x_{1}$ and $x_{2}$.

Due to the scheduling space of the flexible joint manipulator, $\left|x_{1}\right| \leq 2$, and $\left|x_{2}\right| \leq 0.5$ are considered, which result in $|\psi| \leq 2.5$. To partition $\sin (\psi)$ by the extended method for the segregation step, first we start with the most simple PLMs that consist of a single linear model that covers the entire scheduling space, $|\psi| \leq 2.5$. Obviously, this linear model is not an accurate description of $\sin (\psi)$ unless around the origin. Therefore, we split two sides of the scheduling space into two scheduling regimes, $|\psi \pm 1.25| \leq 1.25$. We once again observe that this segregation is not an accurate description of the function, and it needs more regions. Fig. 2 is obtained
by continuing this approach until having the accuracy of $\varepsilon=0.1$. . By an axis-oblique decomposition strategy and using equation (6), the obtained regions according to $x_{1}$ and $x_{2}$ would be as Fig. 3. The upper and lower bounds of each region are given in Table 2.

In addition, for simplicity and having less number of regions with the desired accuracy, the scheduling regions can be reduced to seven regions using extra aggregation. For this purpose, the aggregation of neighboring regions was examined, and were aggregated for the cases with maintained desired accuracy. Therefore, Fig. 4 partitioning is obtained for

Table 2. The upper and lower bounds of $\psi$ after partitioning by segregation method.

| Region No. | Bounds | Region No. | Bounds |
| :--- | :--- | :--- | :--- |
| 1 | $[-2.5000,-1.8750]$ | 6 | $[+0.0000,+0.6250]$ |
| 2 | $[-1.8750,-1.5625]$ | 7 | $[+0.6250,+1.2500]$ |
| 3 | $[-1.5625,-1.2500]$ | 8 | $[+1.2500,+1.5625]$ |
| 4 | $[-1.2500,-0.6250]$ | 9 | $[+1.5625,+1.8750]$ |
| 5 | $[-0.6250,-0.0000]$ | 10 | $[+1.8750,+2.5000]$ |



Fig. 4. Partitioning $\sin (\psi)$ by segregation with extra aggregation.
$\sin (\psi)$. The corresponding regions are shown in Fig. 5. The upper and lower bounds of each region are given in Table 3.

## 3-2- Scheduling Parameters

After decomposing the scheduling space $Z$ in $N$ disjoint regions $Z^{i}$ s, the value of the PLMs parameters (3) consisting of two categories (the set of linear models $\left\{A_{i}, a_{i}\right\} \mathrm{s}$ and the scheduling functions $w_{i}\left(Z^{i}\right)$ s ) should be determined. For a flexible joint manipulator, the number of these parameters reduce as mentioned in (5). They are calculated by linearization of the nonlinear system (2) at the centers of the scheduling regimes $z_{0}^{i} \mathrm{~s}$ for the first category $\left\{A_{i}, a_{i}\right\} \mathrm{s}$, as follows:

$$
\begin{equation*}
A_{i}=\left.\frac{\partial f}{\partial x}\right|_{z_{0}^{i}}, a_{i}=\left.f(x)\right|_{z_{0}^{i}}-A_{i} x ; \quad i=1,2, \ldots, N . \tag{7}
\end{equation*}
$$

For every region of Table 3, the calculated parameters are given in Table 4.

For the second category, the scheduling functions should be determined such that the following properties are satisfied [25]:

$$
\begin{equation*}
w_{i}\left(Z^{i}\right) \geq 0, \sum_{i=1}^{N} w_{i}\left(Z^{i}\right)=1 . \tag{8}
\end{equation*}
$$



Fig. 5. The segregation with extra aggregation regions of $\sin (\psi)$ versus the corresponding $x_{1}$ and $x_{2}$.

Table 3. The upper and lower bounds of $\psi$ after extra aggregation.

| Region No. | Bounds | Region No. | Bounds |
| :--- | :--- | :--- | :--- |
| 1 | $[-2.5,-2.35]$ | 5 | $[0.79,1.57]$ |
| 2 | $[-2.35,-1.57]$ | 6 | $[1.57,2.35]$ |
| 3 | $[-1.57,-0.79]$ | 7 | $[2.35,2.5]$ |
| 4 | $[-0.79,0.79]$ |  |  |

Different selections of $w_{i}\left(Z^{i}\right) \mathrm{s}$ result in different PLMs. If the scheduling functions are selected such that there are overlaps between the regions, the controller design needs to solve a bilinear matrix inequality. Therefore, solving the problem will be much more complex and time-consuming, in addition to the fact that existing methods for solving bilinear matrix inequalities will not necessarily lead to a globally optimal solution. In this paper, to avoid overlaps between the regions, the scheduling functions are defined as follows:

$$
w_{i}\left(Z^{i}\right)=\left\{\begin{array}{ll}
1 & \left(x_{1}, x_{2}\right) \in Z^{i}  \tag{9}\\
0 & \left(x_{1}, x_{2}\right) \notin Z^{i}
\end{array} .\right.
$$

Using (9) and the parameters given in Table 4 in (5), an approximation of the function $\sin \left(x_{1}+x_{2}\right)$ is obtained, as shown in Fig. 6. The difference between these two functions is shown in Fig. 7. It is observed that the maximum value of the differences in the scheduling space is less than 0.1. Using the approximation of this nonlinear term, the PLMs approximation of the whole system (1) is determined easily.

## 4- Output Tracking Control

In this section, we design a controller for the output tracking problem of the flexible joint manipulator. To this end, we use the results of a theorem that is given as follows.

Consider a nonlinear dynamical system and its approximation PLMs as follows:

Table 4. The first category of the parameters for every scheduling regime.

| $i$ | $\left(x_{10}^{i}, x_{20}^{i}\right)$ | $\alpha_{i}$ | $\beta_{i}$ | $\lambda_{i}$ |
| :--- | :--- | :--- | :--- | :--- |
| 1 | $(-1.940,-0.485)$ | -0.7540 | -0.7540 | -2.4854 |
| 2 | $(-1.568,-0.392)$ | -0.3795 | -0.3795 | -1.6689 |
| 3 | $(-0.944,-0.236)$ | +0.3809 | +0.3809 | -0.4751 |
| 4 | $(0,0)$ | +1 | +1 | 0 |
| 5 | $(+0.944,+0.236)$ | +0.3809 | +0.3809 | +0.4751 |
| 6 | $(+1.568,+0.392)$ | -0.3795 | -0.3795 | +1.6689 |
| 7 | $(+1.940,+0.485)$ | -0.7540 | -0.7540 | +2.4854 |



Fig. 6. $\sin \left(x_{1}+x_{2}\right)$ (left-side) and its approximation PLMs (right-side).

$$
\begin{equation*}
\dot{x}=f(x), \tag{10}
\end{equation*}
$$

$$
\begin{equation*}
\dot{x}=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left(A_{i} x+a_{i}\right) . \tag{11}
\end{equation*}
$$

It is assumed that, $f($.$) is a sufficiently smooth func-$ tion. It has been proved that the mismatch between the system and PLMs is bounded by $\left\|F_{i}(x)\right\|_{2} \leq L_{i}\left\|x-x_{0}^{i}\right\|_{2}$, and $L_{i}\left\|x-x_{0}^{i}\right\|_{2} \leq \varepsilon^{2}$, where, $L_{i}$ is a finite positive number and $F_{i}(x)$ is the Taylor expansion remainder of $f(x)$ at the center point $x_{0}^{i}$. For stability of both systems, the following theorem holds:


Fig. 7. Difference between $\sin \left(x_{1}+x_{2}\right)$ and its approximation PLMs.

$$
\begin{aligned}
& \left(\begin{array}{cc}
A_{i}^{T} P+P A_{i}+L_{i} \tau_{i 1} I & P \\
P & -\tau_{i 1} I
\end{array}\right)<0 \\
& \text { if } \quad L_{i} x_{0}^{i T} x_{0}^{i} \leq \varepsilon^{2}, \\
& \left(\begin{array}{ccc}
A_{i}^{T} P+P A_{i}+L_{i}\left(\tau_{i 1}-\tau_{i 2}\right) I & P & P a_{i}+L_{i} x_{0}^{i}\left(\tau_{i 2}-\tau_{i 1}\right) \\
P & -\tau_{i 1} I & 0 \\
a_{i}^{T} P+L_{i} x_{0}^{i T}\left(\tau_{i 2}-\tau_{i 1}\right) & 0 & L_{i} x_{0}^{i T} x_{0}^{i}\left(\tau_{i 1}-\tau_{i 2}\right)+\varepsilon^{2} \tau_{i 2}
\end{array}\right)<0 \text {; } \\
& \text { if } L_{i} x_{0}^{i T} x_{0}^{i}>\varepsilon^{2} .
\end{aligned}
$$

Theorem 1 [25]: The nonlinear dynamics (10) and its $\varepsilon$-accuracy approximation PLMs (11) are asymptotically stable, if there exist a matrix $P=P^{T}>0$, and scalars $\tau_{i j} \geq 0$ with $i \in\{1, \cdots, N\}$ and $j \in\{1,2\}$, such that the following inequalities hold:

We define a new dynamic for output tracking control of the flexible joint manipulator, as follows:

$$
\begin{equation*}
\dot{q} \triangleq r(t)-y(t) \tag{13}
\end{equation*}
$$

where $r(t)$ is the reference signal. Therefore, using (2) and (13), the augmented state space equations of the flexible joint manipulator will be as follows:

$$
\begin{align*}
& \dot{x}=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{A_{i} x+b u+a_{i}\right\}  \tag{14}\\
& \dot{q}=r-c x
\end{align*}
$$

Now, using the state feedback control as:

$$
u(t)=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right) u_{i}(t) ; \quad u_{i}(t)=\left[\begin{array}{ll}
-k_{i} & -m_{i}
\end{array}\right]\left[\begin{array}{l}
x \\
q
\end{array}\right]-\sigma_{i},(15)
$$

where $k_{i}, m_{i}$, and $\sigma_{i}$ are the controller parameters, the closed-loop system becomes:

$$
\begin{aligned}
\dot{x} & =\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{A_{i} x+b \sum_{j=1}^{N} w_{j}\left(Z^{i}\right)\left(-k_{j} x-m_{j} q-\sigma_{j}\right)+a_{i}\right\} \\
\dot{q} & =r-c x
\end{aligned}
$$

Using $\sum_{i=1}^{N} w_{j}\left(Z^{j}\right)=1$ from (8), equation (16) is rewritten as follows:

$$
\begin{align*}
\dot{x} & =\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{\sum_{j=1}^{N} w_{j}\left(Z^{j}\right) A_{i} x+b \sum_{j=1}^{N} w_{j}\left(Z^{j}\right)\left(-k_{j} x-m_{j} q-\sigma_{j}\right)+\sum_{j=1}^{N} w_{j}\left(Z^{j}\right) a_{i}\right\} \\
& =\sum_{i=1}^{N} \sum_{j=1}^{N} w_{i}\left(Z^{i}\right) w_{j}\left(Z^{j}\right)\left\{\left(A_{i}-b k_{j}\right) x-b m_{j} q+\left(a_{i}-b \sigma_{j}\right)\right\},  \tag{17}\\
\dot{q} & =r-c x .
\end{align*}
$$

It is simplified to:

$$
\begin{align*}
\dot{x} & =\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{\left(A_{i}-b k_{i}\right) x-b m_{i} q+\left(a_{i}-b \sigma_{i}\right)\right\}  \tag{18}\\
\dot{q} & =r-c x
\end{align*}
$$

By selecting the scheduling functions as (9), results in:
$w_{i}\left(Z^{i}\right) w_{j}\left(Z^{j}\right)=\left\{\begin{array}{cc}w_{i}\left(Z^{i}\right) & i=j \\ 0 & i \neq j\end{array}\right.$.

Therefore, (18) can be written as:

$$
\left[\begin{array}{l}
\dot{x}  \tag{20}\\
\dot{q}
\end{array}\right]=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{\left(\left[\begin{array}{cc}
A_{i} & 0_{4 \times 4} \\
-c & 0_{1 \times 1}
\end{array}\right]-\left[\begin{array}{c}
b \\
0_{1 \times 1}
\end{array}\right]\left[\begin{array}{ll}
k_{i} & m_{i}
\end{array}\right]\right)\left[\begin{array}{l}
x \\
q
\end{array}\right]+\left[\begin{array}{c}
a_{i} \\
r
\end{array}\right]-\left[\begin{array}{c}
b \\
0_{1 \times 1}
\end{array}\right] \sigma_{i}\right\},
$$

which is rewritten as:

$$
\left[\begin{array}{c}
\dot{x}  \tag{21}\\
\dot{q}
\end{array}\right]=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{\left(\overline{A_{i}}-\bar{b} \bar{k}_{i}\right)\left[\begin{array}{l}
x \\
q
\end{array}\right]+\left(\overline{a_{i}}-\bar{b} \bar{\sigma}_{i}\right)\right\}
$$

By defining following vectors and matrices, we have:

$$
\begin{align*}
& \bar{A}_{i} \triangleq\left[\begin{array}{cc}
A_{i} & 0_{4 \times 4} \\
-c & 0_{1 \times 1}
\end{array}\right], \quad \bar{b} \triangleq\left[\begin{array}{c}
b \\
0_{1 \times 1}
\end{array}\right], \quad \bar{k}_{i} \triangleq\left[\begin{array}{ll}
k_{i} & m_{i}
\end{array}\right], \quad \bar{a}_{i} \triangleq\left[\begin{array}{c}
a_{i} \\
r
\end{array}\right],  \tag{22}\\
& \bar{\sigma}_{i} \triangleq \sigma_{i}
\end{align*}
$$

Equation (21) is described by:

$$
\begin{equation*}
\dot{\bar{x}}=\sum_{i=1}^{N} w_{i}\left(Z^{i}\right)\left\{G_{i} \bar{x}+\eta_{i}\right\} \tag{23}
\end{equation*}
$$

using the following definitions:

$$
\bar{x} \triangleq\left[\begin{array}{l}
x  \tag{24}\\
q
\end{array}\right], \quad G_{i} \triangleq \bar{A}_{i}-\bar{b} \bar{k}_{i}, \quad \eta_{i} \triangleq \bar{a}_{i}-\bar{b} \bar{\sigma}_{i}
$$

It is seen that (23) has the same form as the dynamical equation (11). Therefore, from Theorem 1, asymptotic stability of the closed-loop system (21) is guaranteed if the following conditions hold:

$$
\begin{aligned}
& \left(\begin{array}{cc}
G_{i}^{T} P+P G_{i}+L_{i} \tau_{i 1} I & P \\
P & -\tau_{i 1} I
\end{array}\right)<0 ; \\
& \text { if } \quad L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i} \leq \varepsilon^{2},
\end{aligned}
$$

$$
\left(\begin{array}{ccc}
G_{i}^{T} P+P G_{i}+L_{i}\left(\tau_{i 1}-\tau_{i 2}\right) I & P & P \eta_{i}+L_{i} \bar{x}_{0}^{i}\left(\tau_{i 2}-\tau_{i 1}\right) \\
P & -\tau_{i 1} I & 0 \\
\eta_{i}^{T} P+L_{i} \bar{x}_{0}^{i T}\left(\tau_{i 2}-\tau_{i 1}\right) & 0 & L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i}\left(\tau_{i 1}-\tau_{i 2}\right)+\varepsilon^{2} \tau_{i 2}
\end{array}\right)<0 ;
$$

$$
\text { if } \quad L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i}>\varepsilon^{2}
$$

or equivalently:

$$
\begin{align*}
& \left(\begin{array}{cc}
\left(\overline{A_{i}}-\bar{b} \bar{k}_{k_{i}}\right)^{T} P+P\left(\bar{A}_{i}-\bar{b} \bar{k}_{k_{i}}\right)+L_{i} \tau_{i 1} I & P \\
\quad-\tau_{i l} I
\end{array}\right)<0 ; \\
& \forall L_{i} \bar{x}_{0}^{I T} \bar{x}_{0}^{I} \leq \varepsilon^{2}, \tag{26}
\end{align*}
$$

$\left(\begin{array}{ccc}\left(\overline{A_{i}}-\bar{b} \overline{k_{i}}\right)^{T} P+P\left(\overline{A_{i}}-\bar{b} \bar{k}_{i}\right)+L_{i}\left(\tau_{i 1}-\tau_{i 2}\right) I & P & P\left(\overline{a_{i}}-\bar{b} \bar{\sigma}_{i}\right)+L_{i} \bar{x}_{0}^{\prime}\left(\tau_{i 2}-\tau_{i 1}\right) \\ P & -\tau_{i l} I & 0 \\ \left(\overline{a_{i}}-\bar{b} \bar{\sigma}_{i}\right)^{r} P+L_{i} \bar{x}_{0}^{r}\left(\tau_{12}-\tau_{i 1}\right) & 0 & L_{i} \bar{x}_{0}^{T} \bar{x}_{0}^{T}\left(\tau_{1}-\tau_{i 2}\right)+\varepsilon^{2} \tau_{12}\end{array}\right)<0 ;$ $\forall L_{i} \bar{x}_{0}^{i \pi} \bar{x}_{0}^{i}>\varepsilon^{2}$.

Defining new variables as follows:

$$
\begin{equation*}
P \bar{b} \bar{\sigma}_{i} \triangleq \mu_{i}, \quad P \bar{b} \bar{k}_{i} \triangleq Y_{i} \tag{27}
\end{equation*}
$$

The conditions (26) become as linear matrix inequalities:

$$
\begin{align*}
& \left(\begin{array}{cc}
\bar{A}_{i}^{T} P+P \bar{A}_{i}-\left(Y_{i}+Y_{i}^{T}\right)+L_{i} \tau_{i 1} I & P \\
P & -\tau_{i 1} I
\end{array}\right)<0 \\
& \forall L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i} \leq \varepsilon^{2},  \tag{28}\\
& \left(\begin{array}{ccc}
\bar{A}_{i}^{T} P+P \bar{A}_{i}-\left(Y_{i}+Y_{i}^{T}\right)+L_{i} \tau_{i 1} I-L_{i} \tau_{i 2} I & P & \left.P \bar{a}_{i}-\mu_{i}+L_{i} \bar{x}_{0}^{i} \tau_{i 2}-L_{i} \bar{x}_{0}^{i} \tau_{i 1}\right) \\
P & -\tau_{i 1} I & 0 \\
\bar{a}_{i}^{T} P-\mu_{i}^{T}+L_{i} \bar{x}_{0}^{i T} \tau_{i 2}-L_{i} \bar{x}_{0}^{i T} \tau_{i 1} & 0 & L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i} \tau_{i 1}-L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i} \tau_{i 2}+\varepsilon^{2} \tau_{i 2}
\end{array}\right)<0 ; \\
& \forall L_{i} \bar{x}_{0}^{i T} \bar{x}_{0}^{i}>\varepsilon^{2} .
\end{align*}
$$

Solving (28) for $P, \mu_{i}$, and $Y_{i}$, and substituting them into (27), the controller parameters, $\bar{k}_{i}=\left[k_{i}, m_{i}\right]$ and $\bar{\sigma}_{i}=\sigma_{i}$ are obtained. These parameters were obtained for the flexible joint manipulator, as given in Table 5.

Applying the controller to the nonlinear system (1), it is seen that the output $y=\theta+\alpha$, tracks the reference signal well, which is shown in Fig. 8. The reference signal has been considered as a combination of functions such as step, ramp, and sinusoidal in all operating spaces of the system, to test the performance of the controller in the various regions. As a comparison, we also used state feedback control in a single linear model that the tracking result of which is also shown in Figure 8. It is observed that better tracking is achieved using PLMs due to linearization around several operating points that lead to a better approximation of the model compared to the linearized model around only one operating point. In a single linear model, the greater the distance from the operating point, the greater the error. Furthermore, Fig. 9 shows the control signal for the PLMs method. As seen in Fig. 10, a good approximation is achieved for the state variables of the nonlinear system and its approximated system PLMs using the same control signal.

Table 5. Controller parameters.

| i | $k_{1 i}$ | $k_{2 i}$ | $k_{3 i}$ | $k_{4 i}$ | $m_{i}$ | $\sigma_{i}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | -0.0055 | +3.1230 | +13.4750 | -13.9163 | -10.5281 | 0 |
| 2 | +0.0006 | +3.1235 | +13.4819 | -13.9244 | -13.1601 | 0 |
| 3 | +0.0117 | +3.1251 | +13.4952 | -13.9403 | -10.9050 | 0 |
|  |  |  |  |  |  |  |
|  | +0.0196 | +3.1288 | +13.5077 | -13.9557 | -11.1588 | 0 |
| 5 | +0.0117 | +3.1251 | +13.4952 | -13.9403 | -10.9050 | 0 |
| 6 | +0.0006 | +3.1235 | +13.4819 | -13.9244 | -13.1601 | 0 |
| 7 | -0.0055 | +3.1230 | +13.4750 | -13.9163 | -10.5281 | 0 |



Fig. 8. The output tracking by two modeling methods; PLMs and a linear model.


Fig. 9. Control signal.





Fig. 10. State variables of the nonlinear system and its PLMs approximation.

## 5-Conclusion

This study has investigated the output tracking control problem of a single-link flexible joint manipulator. Due to the existence of powerful tools in linear control theory, the PLMs based methods were applied for modeling and controlling the flexible joint manipulator. In order to decrease the number of local models and to avoid the curse of dimensionality, an axis oblique decomposition was used for the partitioning purpose. To determine the parameters of each region, the scheduling functions were selected such that the interference between the regions and therefore requiring to solve bilinear matrix inequality were avoided. Hence, by solving an LMI, the controller was designed and applied to the flexible dynamics.

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